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Area Efficient Multiband Frequency Divider

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ABSTRACT:In this paper the prescaler circuit which is used by frequency synthesizers of Bluetooth, zigbee and WLAN is proposed with multi modulus 32/33/47/48 prescaler, New E-TSPC 2/3 prescaler, P-counter and S-counter. This proposed prescaler can divide the frequency in three bands 0f 2.4-2.484GHz, 5.15-5.35GHz and 5.725-5.825GHz with a resolution selectable from 1-25MHz.The Area and power consumed by the multimodulusprescaler circuit is minimized.

KEYWORDS:DFF, frequency synthesizer, E-TSPC, wireless LAN (WLAN), true single phase clock (TSPC)

1. INTRODUCTION

Wireless LAN (WLAN) in the multigigahertz bands, such as HiperLAN II and IEEE 802.11a/b/g, are the leading standards for high-rate data transmissions, and standards like IEEE 802.15.4 are recognized for low-rate data transmissions. The demand for lower cost, lower power, and multiband RF circuits increased in conjunction with need of higher level of integration. The frequency synthesizer, usually implemented by a phase-locked loop (PLL) is one of the power-hungry blocks in the RF front-end, and the first-stage frequency divider consumes a large portion of power in a frequency synthesizer. The integrated synthesizers for WLAN applications at 5GHz reported in [3] consume up to 25mW in CMOS realizations.

The frequency synthesizer reported in [4] uses an E-TSPC prescaler as the first-stage divider, but the divider consumes around 6.25mW. A low-power clock multiband divider for Bluetooth, Zigbee, and IEEE 802.15.4 and 802.11 a/b/g WLAN frequency synthesizers based on pulse-swallow topology is reported in [1], and it is implemented using a 0.18µm CMOS technology. The multiband divider consists of a wide band multimodulus 32/33/47/48 prescaler and swallow (S) counter and p-counter. In this proposed system the E-TSPC 2/3 prescaler in [1] is replaced by the New E-TSPC 2/3 prescaler[2], which consumes low power and area when compared to the existing design.

II. BLOCK DIAGRAM

In the existing system, a dynamic logic multiband flexible integer-N divider based on pulse-swallow topology is proposed, which uses a low-power wide band 2/3 prescaler and a wide band multimodulus 32/33/47/48 prescaler as shown in Fig.1. The divider also used an improved low-power loadable bit-cell for the Swallow S-counter and P-Counter. Fig.2 shows the proposed multi modulus 23/33/47/48 prescaler block with New E-TSPC 2/3 Prescaler [2]



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Fig.2. Proposed Multi-Modulus 32/33/47/48 Prescaler

III. NEW E-TSPC 2/3 PRESCALER

This New low power E-TSPC 2/3 prescaler is introduced from [2] which is shown in Fig.3, and this is the further improved version of the Existing prescaler. It consist of two flipflops and the and gate as like previous design. The OR gate used for the divide control is replaced with a PMOS transistor, it act likes a switch. Note that there is a negation bubble at one of the AND gate's input. The output Q of FF1 is thus complemented before being fed to FF2. When the switch is open, the input from FF1 is disconnected and FF2 alone divides the clock frequency by 2. When the switch is close, FF1 and FF2 are linked to form a counter with three distinct states. Besides the speed advantage, E-TSPC FFs are particularly useful for low voltage operations because of the minimum height in transistor stacking.





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Other than the two E-TSPC FFs, only one pMOS Transistor (P_{MC}) is needed. The pMOS transistor controlled by the divide control signal serves as the switch. The AND gate plus its input inverter are achieved by way of wired-AND logic using no extra transistors at all. The proposed design scheme is far more sophisticated than the measure of simply adding one pass transistor may suggest. First of all, unlike any previous designs, the E-TSPC FF design remains intact without any logic embedding. Both speed and power behaviors are not affected, which indicates a performance edge over the logic embedded FF design. Secondly, the inverter to complement the one of the two E-TSPC FF outputs for divide-by-3 operations is removed in the proposed design. The circuit simplification, again, suggests the improvements in both speed and power performances.

IV. 32/33/47/48 PRESCALER

The proposed wide band multimodulusprescaler is shown in Fig.4, which can divide the input frequency by 32, 33, 47, and 48. It is similar to the 32/33 prescaler. But it consists of additional inverter and a multiplexer as like existing design. This prescaler performs additional divisions (divide- by-47 and divide-by-48) without any extra flip-flop, thus saving a considerable amount of power and also reducing the complexity of multiband divider which will be discussed below. The multimodulusprescaler consists of New E-TSPC 2/3 prescaler [2], four divide-by-2 circuits ((AD)=16) and combinational logic circuits for getting multiple division ratios. Beside the usual MOD signal for controlling N/(N+1)) divisions, the additional control signal Sel is used to switch the prescaler between 32/33 and 47/48 modes.



Fig.4.Multimodulus 32/33/47/48 PrescalerUsing New E-Tspc 2/3 Prescaler

a. CASE 1: Sel= '0'

When Sel = '0', the output from the NAND2 gate is directly transferred to the input of 2/3 prescaler and the multimodulusprescaler operates as the normal 32/33 prescaler, where the division ratio is controlled by the logic signal MOD. If MC='1', the 2/3 prescaler operates in the divide-by-2 mode and when MC='0', the 2/3 prescaler operates in the divide-by-3 mode. If MOD = '1', the NAND2 gate output switches to logic "1" (MC = '1') and the wide band prescaler operates in the divide- by-2 mode for entire operation. The division ratio N of the multimodulusprescaler is $N = (AD^*N_I) + (0^*(N_I + I)) = 32.$ (1)

Where $N_1 = 2^{\circ}$ and $AD = 16^{\circ}$ is fixed for the entire design. If $MOD = 0^{\circ}$, for 30 input clock cycles 'MC' remains at logic "1", where wideband prescaler operates in divide-by-2 mode and, for three input clock cycles, 'MC' remains at logic "0" where the wideband prescaler operates in the divide-by-3 mode. The division ratio N+1 of themultimodulus prescaler is

$$N+1 = ((AD - 1) * N_1) + (1 * (N_1 + 1)) = 33.$$
(2)

4.2 CASE 2: Sel = '1'

When Sel = '1', the inverted output of the NAND2 gate is directly send to the input of 2/3 prescaler and the multimodulusprescaler operates as a 47/48 prescaler, and the division ratio is controlled by the signal 'MOD'. If MC='1', the 2/3 prescaler operates in divide-by-3 mode and when MC='0', the 2/3 prescaler operates in divide-by-2 mode'. If MOD='1', the division ratio N +1 performed by the multimodulusprescaler is same as (1) except that the wide band prescaler operates in the divide-by-3 mode for the entire operation given by

$$N+1 = (AD * (N_1 + 1)) + (0 * N_1) = 48$$
(3)
If MOD = '1', the division ratio N performed by the multimodulusprescaler is

$$N = ((AD - 1) * (N_{1+1})) + (1 * N_1) = 47$$
(4)



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V. MULTIBAND FLEXIBLE DIVIDER

The single-phase clock multiband flexible divider which is shown in Fig.1, consists of the Proposed multimodulus 32/33/47/48 prescaler, a 7-bit programmable P-counter and a 6-bit swallow S-counter. The multimodulus 32/33/47/48 precaler is briefly discussed in Section IV. The control signal Sel decides whether the divider is operating in lower frequency band or higher band. The operation of P-counter and S-counter are same as the existing system[1]

VI. SIMULATIONS

The proposed multimodulus prescaler has the maximum operating frequency of 7.2 GHz (simulation)[1]. The simulation results of this proposed multimodulus prescaler in different mode of operation (32/33/47/48) is shown in the Fig.7. These simulation waveforms are carried out by using Modelsim6.4c(Verification tool)

Messag /dock1/dk /dock1/md /dock1/sel /dock1/fout /dock1/count	yes 1 1 0 1 34		2 <u>)</u> 3 <u>)</u> 4 <u>(</u>		9 <u>(10)(11)</u> (a	12)13)14)15 .)	<u>, , , , , , , , , , , , , , , , , , , </u>	20 <u>/21 /22 /23 /</u>	24 <u>25)26)27)</u>	28) 29 (30) 31)	<u>111</u>
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(c)



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(d)

Fig.7. Simulation Result of MultimodolusPrescaler during (a)divide by 32 mode, (b) divide by 33 mode, (C) divide by 47 mode, (d) divide by 48 mode.

The simulation waveform of the entire divider block including P& S counter and "FD" for some clock cycles is shown in the Fig. 8.



Fig.8. Simulation Waveform of Multiband Divider

The proposed design is implemented using Xilinx ISE 10.1i (Synthesis Tool). The area utilization (Equivalent gate count) of the proposed multiband divider using New E-TSPC 2/3 prescaler is less, when compared to the existing design [1], without any change in the operation sequence. The area utilization of proposed and existing system is shown in the Fig.9. The result comparison are shown in the Table 1



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Device Utilization Summary				
Logic Utilization	Used	Available	Utilization	
Total Number Slice Registers	29	6,144	1%	
Number used as Flip Flops	19			
Number used as Latches	10			
Number of 4 input LUTs	51	6,144	1%	
Logic Distribution				
Number of occupied Slices	51	3,072	1%	
Number of Slices containing only related logic	45	45	100%	
Number of Slices containing unrelated logic	0	45	0%	
Total Number 4 input LUTs	62	6,144	1%	
Number used as logic	51			
Number used as Shift registers	1			
Number of bonded <u>IOBs</u>	12	178	6%	
Number of GCLKs	1	4	25%	
Number of GCLKIOBs	1	4	25%	
Total equivalent gate count for design	636			
Additional JTAG gate count for IOBs	624			

(a)

Device Utilization Summary				
Logic Utilization	Used	Available	Utilization	
Total Number Slice Registers	29	6,144	1%	
Number used as Flip Flops	19			
Number used as Latches	10			
Number of 4 input LUTs	51	6,144	1%	
Logic Distribution				
Number of occupied Slices	45	3,072	1%	
Number of Slices containing only related logic	46	46	100%	
Number of Slices containing unrelated logic	0	46	0%	
Total Number of 4 input LUTs	51	6,144	1%	
Number of bonded <u>IOBs</u>	12	178	6%	
Number of GCLKs	1	4	25%	
Number of GCLKIOBs	1	4	25%	
Total equivalent gate count for design	508			
Additional JTAG gate count for IOBs	624			

(b)

Fig.9. (a) Area Utilization of Existing Design, (b) Area Utilization of Proposed Design.

Table 1 Result Comparison.

Logic Utilization	In This Paper	Existing Design [1]
Total No. of Slice Registers	29	29
No. of Occupied Slice	45	51
Total No. of 4 Input LUTs	51	62
Total Equivalent Gate Count	508	636



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Fig.10. Comparison chart

VII. CONCLUSION

In this paper, New E-TSPC 2/3 prescaler is used in wide band multimodulus 32/33/47/48 prescaler. A dynamic logic multiband flexible integer-N divider is designed which uses E-TSPC 2/3 prescaler [2]. The P-counter and S-counter values are programmable and it is programmed to the divider between the range of 1 to 6.2 GHz with finest resolution of 1 MHz and variable channel spacing. since interest lies in the 2.4- and 5–5.825-GHz bands of operation, the P- and S-counters are programmed accordingly. The proposed multiband flexible divider also uses an improved loadable bit-cell for P & S counters. The overall equivalent gate count of the design is reduced to 508 from 636.

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